Chebyshev Filter at 197.12 MHz Frequency for Radar System

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Abstract: This paper describes the design and simulation of bandpass filter at 197.12 MHz frequency cut off. which use Chebyshev methode, Chebyshev Filter is allowing ripple on the frequency response. But this type of filter is easy for tunning at cut off frequency. This Filter is used for Radar system, which have small bandwidth, it's around 4 MHz, and also compatible to baseband frequency. In order to design this filter, first of all, we calculate the value of inductor and capasitor, and then simulate it with software. The design filter was simulated using Elsie Tonne version 2.4. From the simulation results, we can see that the VSWR is 1.0021, return loss -59.74 dB.

Keywords – Filter, Chebyshev, bandwidth, insertion loss, vswr,

I. INTRODUCTION

Every Project in Radar Communication needs special specification for each component which can't find in the market. One of this component is Band Pass Filter at 197.12 MHz, which use Chebyshev method. Chebyshev method used to make special specification because the order of the filter can be higher than Butterworth. Bandpass filter is combination of low-pass filter and high-pass filter. So the bandwidth of this filter is between upper and lower cut off frequencies, it can be measured at half power points (gain -3dB).

The "ideal" Bandpass filter is no ripple and attenuate all frequencies outside the passband. But sometime we need Bandpass Filter which have some ripple on the Frequency respons, because we can tune the cut off frequency more easier. However the filter which use this type is more difficult to pabricated.

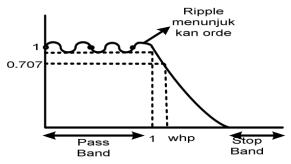


Figure 1. Respon Frequency of Chebyshev Filter

Chebysev filter have ripple in passband and stopband. From the figure we can see the respon frequency of chebysev which has ripple in passband. This filter used to separate one band of frequencies from another.

II. FILTER DESIGN

First of all for this design we have to calculate the element value of series capasitor , series inductor, parallel capasitor and parallel inductor as follows :

$$L_{series} = \frac{\left(\frac{R_0}{(f_c_2 - f_{c1})\pi}\right)}{2} \tag{1}$$

$$C_{series} = 2\left(\frac{f_{c2}-f_{c1}}{R_0(f_{c2},f_{c1})4\pi}\right)$$
(2)

$$L_{parallel} = 2\left(\frac{R_0(f_{c2}-f_{c1})}{(f_{c2}.f_{c1})4\pi}\right)$$
(3)

$$C_{prallel} = \frac{\left(\frac{1}{R_0(f_{c2} - f_{c1})\pi}\right)}{2} \tag{4}$$

From the formula we can calculate element value of component and put to the figure 4. So the design filter can be simulate with elsie Tonne.

According to formula (1), (2), (3), and (4) we can calculate series capasitor , series inductor, parallel capasitor and parallel inductor with some of parameter such as lower frequency cut off (fc_l) , upper frequency cut off (fc_2) , Impedance input (R_o), Bandwidth (BW).

 $fc_l = 195.12 \text{ MHz} = 195.12 \text{ x} 10^6 \text{ Hz}$ $fc_2 = 199.12 \text{ MHz} = 199.12 \text{ x} 10^6 \text{ Hz}$ $BW = fc_2 - fc_l = 4 \text{ MHz}$ $R_0 = 50 \Omega$

To find the value of

1. $L_{series} =$

$$L_{series} = \frac{\left(\frac{R_0}{(f_{c2} - f_{c1})\pi}\right)}{2}$$
$$L_{series} = \frac{\left(\frac{50}{(195.12 \ x \ 10^6 - 199.12 \ x \ 10^6)\frac{22}{7}}\right)}{2} = 1,988 \ uH$$

2. $C_{series} =$

$$C_{series} = 2\left(\frac{f_{c2} - f_{c1}}{R_0(f_{c2}, f_{c1})4\pi}\right)$$

$$C_{series} = 2\left(\frac{(199.12 - 195.12)10^6}{50((199.12 \times 10^6)x (195.12 \times 10^6))4\frac{22}{7}}\right) = 0.3275809 \ pF_{c2}$$

3. $L_{parallel} =$

$$L_{parallel} = 2\left(\frac{R_0(f_{c2} - f_{c1})}{(f_{c2}, f_{c1})4\pi}\right)$$
$$L_{parallel} = 2\left(\frac{50((199.12 \times 10^6) - (195.12 \times 10^6))}{((199.12 \times 10^6) \times (195.12 \times 10^6))4\frac{22}{7}}\right) = 0,8189524 \, nH$$

4. C_{parallel} =

$$C_{prallel} = \frac{\left(\frac{1}{R_0(f_{c2} - f_{c1})\pi}\right)}{2}$$

$$C_{prallel} = \frac{\left(\frac{1}{50\left((199.12 \times 10^{6}) - (195.12 \times 10^{6})\right)\frac{22}{7}}\right)}{2} = 0,795454 \, nH$$

III. SIMULATION

Before realization the filter based on the element value from calculation with formula (1), (2), (3) and (4). We can simulate the element value on the software simulation which called Elsie Tonne 2.4 version. Some parameters needs to design such as the type of filter (Chebysev or Butterworth), Bandwidth, order etc, as can see in the figure 2.

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Capacitor-input highpass Cinductor-input highpass Series-input bandstop Shuml-input bandstop ?						See normalized values				0.0001 VSWR 1.0096 Return: -46.378 Center Freq (Fo) 197.12M				
					Cauer even terms © Best selective/ © Equal terms				Stopband width (Fs)					
Dimensions Plot linewidth:					Cauer BPF topology R Normal C 2/g-zag				Stopband Depth (dB) (As)					

Figure 2. The parameter of filter design

We choose topology from the filter use shunt-input bandpass with compare the element value in the market. Other parameters are needed to analysis such as start and stop frequency, VSWR (Voltage Standing Wave Ratio), Q of inductor and Q of capasitor etc, as can see in figure 3.

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Figure 3. Analysis Parameters

The frequency response from simulation, we get the response wich no ripple and have steepness of the response curve as can see in figure 4.

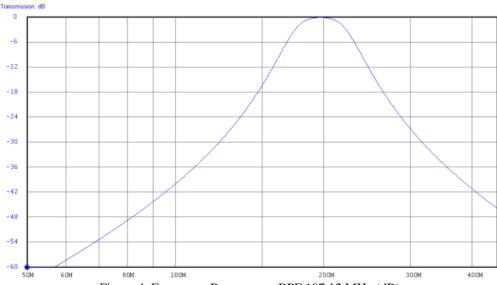


Figure 4. Frequency Response at BPF 197.12 MHz (dB)

The Phase on 197.12 MHz is 0 degree, it means the input signal and output signal have a same position wave degree. We also can see the result in the figure 5.

Frequency, Transmis	sion, Phase,	Return,	Delay,	VSWR:	Zin,	Zangle,	
Hz: dB:	degrees:	dB:	seconds:		ohms:	degrees:	
Hz: dB: 187.05M -10.3 187.46M -9.3046 187.87M -8.3027 188.28M -7.3036 188.69M -6.319 189.1M -5.3646 189.52M -4.459 189.93M -3.6226 190.35M -2.8753 191.18M -1.7047 191.6M -1.2896 192.02M -0.9791 192.44M -0.7573 193.28M -0.6059 193.7M -0.4443	degrees: -1.355 -6.075 -11.279 -17.019 -23.338 -30.266 -37.806 -45.923 -54.537 -63.517 -72.701 -81.913 8 -90.998 8 -99.839 7 -108.37 8 -116.55 5 -124.41 6 -131.97	dB: -0.69251 -0.84477 -1.041 -1.295 -1.6244 -2.0514 -2.6023 -3.3066 -4.1954 -5.2982 -6.6401 -8.242 -10.121 -12.299 -14.804 -17.688 -21.041 -25.021	seconds: 29.18n 32.05n 35.27n 38.81n 42.63n 46.64n 50.65n 54.41n 57.61n 59.93n 61.15n 61.21n 60.23n 58.49n 56.29n 53.93n 51.65n 49.58n	25.098 20.58 16.708 13.44 10.725 8.5076 6.7255 5.3169 4.2208 3.3799 2.7424 2.2635 1.9063 1.641 1.4447 1.3002 1.1947 1.1189	ohms: 48.896 52.979 57.881 63.847 71.196 80.298 91.438 104.33 116.97 124.41 121.6 109.78 95.097 81.927 71.605 64.011 58.642 54.988	degrees: 85.436 84.427 83.076 81.232 78.666 75.025 69.783 62.218 51.635 38.162 23.781 11.492 2.9191 -2.1097 -4.4667 -5.027 -4.4971 -3.4173	*
194.55M -0.3826 194.97M -0.3681 195.4M -0.3586 195.83M -0.3519 196.26M -0.3470 196.69M -0.3433	3 -146.36 6 -153.28 9 -160.07 4 -166.76	-29.936 -36.491 -46.959 -52.885 -48.676 -51.196	47.79n 46.29n 45.08n 44.12n 43.37n 42.81n	1.0658 1.0304 1.009 1.0045 1.0074 1.0055	52.619 51.185 50.401 50.039 49.917 49.9	-2.1848 -1.0707 -0.23498 0.25612 0.41115 0.29401	
197.12M -0.3408		-59.74	42.4n	1.0021	49.897	0.00547	
197.55M -0.3394 197.98M -0.3391 198.41M -0.3398 198.85M -0.3416 199.28M -0.3449 199.28M 0.2603	1 167 3 160.46 6 153.89 7 147.25	-49.9 -45.115 -44.011 -46.882 -49.809	42.12n 41.98n 41.96n 42.09n 42.38n	1.0064 1.0112 1.0127 1.0091 1.0065	49.869 49.823 49.812 49.935 50.323 51.144	-0.33427 -0.60237 -0.68921 -0.51342 -0.03613	Ш

Figure 5. The Tabulate Result of the design

For the VSWR in 197.12 MHz is 1.0021, so no reflected wave. It's indicate the input signal is wholly transferred by the filter 100%, the transmission is 0 dB and the return loss is more than -59.74 dB.

IV. PABRICATION

The Pabrication of this filter is use PCB duroid 5870, which has Dielectric constant is around 2.33, and the cut off of frequency is less than 1 GHz.

Using the Trimmer Capasitor is make us easy to find the value of capasitor because from the calculation result the value of the capasitor is not available in the market.

For the Inductor, we use the SMT Inductor to get the smalles dimention and appropriate the value that we needed.

After we made this filter, we can measure the characteristic of this filter with Vector Network Analyzer and the result is show in the fig. 7. we can see that the measurement result for this filter is approximately same with the simulation.

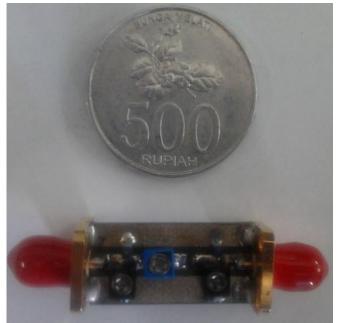


Figure 6. The Pabrication of 197.12 MHz Bandpass Filter

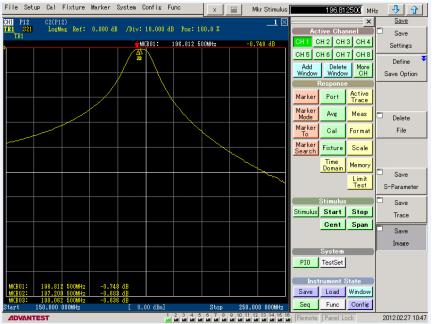


Figure 7. The Measurement of this Filter

For design the filter at the frequency 97.12 MHz we need to calculate the elements value with formula and we have to know about the components value in the market so we can approximately the values which we need and input the values in the parameters culomn at software simulation. In this design we get the good result as we know the frequency response of this filter have ripple because this design use Chebyshev type and use the third order to get steepness for the curve. For the transmission response the curve at the 197.12 MHz have 0.748 dB is approximately equal to 4 MHz bandwidth. The phase of the wave is 0 degree, at the 97.12 MHz and the VSWR is 1.0021, it's represent there are no reflection signal so the input signal is have same ratio with output signal, and the return loss is -59.74 dB, it's means no interference for this filter. From all result parameters are indicate the filter that we design have a good performance and also the pabrication.

Conclusion

The result of the simulation show that the calculation that we made is approximately with the real value of the capasitor and inductor. The design from simulation we can pabricated to get the real Bandpass filter, which has small dimention and good result for the measurement.

Acknowledgements

The author would like to thank to all my colleagues who have assistance on designed and discussed about the filter in this paper.

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